

# ECE304 Homework #1 Solution

## Problem 4.6

a).

$$C_{ox} = \frac{\epsilon_{ox}}{t_{ox}} = \frac{3.45 \times 10^{-11}}{15 \times 10^{-9}} = 2.3 \times 10^{-3} \text{ F/m}^2 = 2.3 \text{ fF}/\mu\text{m}^2$$

$$K'_n = \mu_n C_{ox} = 550 \times 10^{-4} \times 2.3 \times 10^{-3} = 126.5 \mu\text{A}/\text{V}^2$$

b).

$$i_D = \frac{1}{2} k'_n \frac{W}{L} (V_{GS} - V_t)^2 \Rightarrow 100 = \frac{1}{2} \times 126.5 \times \frac{16}{0.8} (V_{GS} - 0.7)^2$$

$$V_{GS} - 0.7 = 0.28 \Rightarrow V_{OV} = 0.28 \text{ V}, V_{GS} = 0.98 \text{ V}$$

$$V_{DSmin} = V_{GS} - V_t = 0.28 \text{ V}$$

c).

For small  $V_{DS}$  (triode region):

$$i_D = k'_n \frac{W}{L} V_{OV} \cdot V_{DS}$$

$$r_{DS} = \frac{V_{DS}}{i_D} = \frac{1}{k'_n \frac{W}{L} V_{OV}} = \frac{1}{126.5 \times 10^{-6} \times \frac{16}{0.8} V_{OV}} = 1000$$

$$\Rightarrow V_{OV} = 0.4 \text{ V}$$

$$\Rightarrow V_{GS} = V_{OV} + V_t = 0.4 + 0.7 = 1.1 \text{ V}$$

## Problem 4.7

$$k'_n = \mu_n C_{ox} = \mu_n \frac{\epsilon_{ox}}{t_{ox}} = 650 \times 10^{-4} \times \frac{3.45 \times 10^{-11}}{20 \times 10^{-9}} = 112.1 \mu\text{A}/\text{V}^2$$

a).  $V_{DS} < V_{GS} - V_t$ : triode region

$$i_D = k'_n \frac{W}{L} \left[ (V_{GS} - V_t) V_{DS} - \frac{1}{2} V_{DS}^2 \right]$$

$$i_D = 112.1 \times 10^{-6} \times 10 \times \left[ (5 - 0.8) \times 1 - \frac{1}{2} \times 1^2 \right] = 4.15 \text{ mA}$$

b).  $V_{DS} = V_{GS} - V_t$ : edge of saturation

$$i_D = \frac{1}{2} k'_n \frac{W}{L} (V_{GS} - V_t)^2$$

$$i_D = \frac{1}{2} \times 112.1 \times 10^{-6} \times 10 \times (2 - 0.8)^2 = 0.8 \text{ mA}$$

c).  $V_{DS} < V_{GS} - V_t$ : triode region

$$i_D = k'_n \frac{W}{L} \left[ (V_{GS} - V_t) V_{DS} - \frac{1}{2} V_{DS}^2 \right]$$

$$i_D = 112.1 \times 10^{-6} \times 10 \times \left[ (5 - 0.8) \times 0.2 - \frac{1}{2} \times 0.2^2 \right] = 0.92 \text{ mA}$$

d).  $V_{DS} > V_{GS} - V_t$ : saturation region

$$i_D = \frac{1}{2} k'_n \frac{W}{L} (V_{GS} - V_t)^2$$

$$i_D = \frac{1}{2} \times 112.1 \times 10^{-6} \times 10 \times (5 - 0.8)^2 = 9.9 \text{ mA}$$

### Problem 4.31:

a). Note that  $[(\partial i_D / i_D) / \partial T] = [(\partial i_D / \partial T) / i_D]$

$$i_D = \frac{1}{2} k'_n \frac{W}{L} (V_{GS} - V_t)^2$$

$$\frac{\partial i_D}{\partial T} = \frac{1}{2} \frac{\partial k'_n}{\partial T} \frac{W}{L} (V_{GS} - V_t)^2 + \frac{1}{2} k'_n \frac{W}{L} \left[ 2 (V_{GS} - V_t) \left( -\frac{\partial V_t}{\partial T} \right) \right]$$

$$\frac{\partial i_D}{\partial T} = \frac{1}{2} \frac{\partial k'_n}{\partial T} \frac{W}{L} (V_{GS} - V_t)^2 - k'_n \frac{W}{L} (V_{GS} - V_t) \frac{\partial V_t}{\partial T}$$

$$[(\partial i_D / i_D) / \partial T] = [(\partial i_D / \partial T) / i_D] = \frac{\partial k'_n / k'_n}{\partial T} - 2 \frac{\partial V_t}{\partial T} \frac{1}{(V_{GS} - V_t)}$$

b). We have

$$\frac{\partial V_t}{\partial T} = -0.002 \text{ V}/^\circ\text{C}, \quad \frac{(\partial i_D / i_D)}{\partial T} = -0.2\% / ^\circ\text{C} = -0.002 / ^\circ\text{C}$$

$$\Rightarrow -0.002 = \frac{(\partial k'_n / k'_n)}{\partial T} - 2 \times (-0.002) \times \frac{1}{5 - 1}$$

$$\Rightarrow \frac{(\partial k'_n / k'_n)}{\partial T} = -0.003 / ^\circ\text{C} = -0.3\% / ^\circ\text{C}$$

### Problem 4.32

Case	Transistor	$V_S$	$V_G$	$V_D$	$I_D$	Type	Mode	$\mu C_{ox} W / L$	$V_t$
a	1	0	2	5	100	N	Saturation	200	1
	1	0	3	5	400	N	Saturation	200	1
b	2	5	3	-4.5	50	P	Saturation	400	-1.5
	2	5	2	-0.5	450	P	Saturation	400	-1.5
c	3	5	3	4	200	P	Saturation	400	-1
	3	5	2	0	800	P	Saturation	400	-1
d	4	-2	0	0	72	N	Saturation	100	0.8
	4	-4	0	-3	270	N	Triode	100	0.8

Case a).

transistor 1:  $V_{GS} = 2\text{ V}$ ,  $V_{DS} = 5\text{ V}$ ,  $I_D = 100\text{ }\mu\text{A}$ . This is an NMOS in saturation region.

$$I_{D1} = 100 = \frac{1}{2}\mu_n C_{ox} \frac{W}{L} (2 - V_t)^2$$

When  $V_{GS} = 3\text{ V}$ ,  $I_D = 400\text{ }\mu\text{A}$ . Similarly we have:

$$\begin{aligned} I_{D2} = 400 &= \frac{1}{2}\mu_n C_{ox} \frac{W}{L} (3 - V_t)^2 \\ \Rightarrow \frac{I_{D1}}{I_{D2}} = \frac{100}{400} &= \frac{(2 - V_t)^2}{(3 - V_t)^2} \\ \Rightarrow V_t = 1\text{ V} &\Rightarrow \mu_n C_{ox} \frac{W}{L} = 200\text{ }\mu\text{A/V}^2 \end{aligned}$$

Case b).

transistor 2:  $V_{GS} = 3 - 5 = -2\text{ V}$ ,  $V_{DS} = -9.5\text{ V}$ . This is a PMOS in saturation region.

$$\begin{aligned} I_{D1} = 50 &= \frac{1}{2}\mu_p \frac{W}{L} (-2 - V_t)^2 \\ I_{D2} = 450 &= \frac{1}{2}\mu_p \frac{W}{L} (-3 - V_t)^2 \\ \Rightarrow \frac{I_{D1}}{I_{D2}} = \frac{50}{450} &= \frac{(2 + V_t)^2}{(3 + V_t)^2} \\ \Rightarrow V_t = -1.5\text{ V} &\Rightarrow \mu_p C_{ox} \frac{W}{L} = 400\text{ }\mu\text{A/V}^2 \end{aligned}$$

Case c).

transistor 3:  $V_{GS} = -2\text{ V}$ ,  $V_{DS} = -1\text{ V}$ . This is a PMOS and it can be either in saturation region or triode region. We assume it is in saturation region:

$$\begin{aligned} I_{D1} = 200 &= \frac{1}{2}\mu_p \frac{W}{L} (-2 - V_t)^2 \\ I_{D2} = 800 &= \frac{1}{2}\mu_p \frac{W}{L} (-3 - V_t)^2 \\ \Rightarrow \frac{I_{D1}}{I_{D2}} = \frac{200}{800} &= \frac{(2 + V_t)^2}{(3 + V_t)^2} \\ \Rightarrow V_t = -1\text{ V} &\Rightarrow \mu_p C_{ox} \frac{W}{L} = 400\text{ }\mu\text{A/V}^2 \end{aligned}$$

The assumption is right, therefore we have:

$$\begin{aligned} V_{DS} = -1\text{ V}, V_{GS} - V_t &= -2 + 1 = -1\text{ V} \\ \Rightarrow V_{DS} = V_{GS} - V_t, &\text{ edge of saturation} \end{aligned}$$

Case d).

transistor 4:  $V_{GS} = 2\text{ V}$ ,  $V_{DS} = 2\text{ V}$ . This is an NMOS in saturation region.

$$I_{D1} = 72 = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (2 - V_t)^2$$

When  $V_{GS} = 4\text{ V}$ ,  $V_{DS} = 1\text{ V}$ , this device is in triode region (considering that  $V_t < 2\text{ V}$ , we have  $V_{DS} < V_{GS} - V_t$ ).

$$I_{D2} = 270 = \mu_n C_{ox} \frac{W}{L} \left[ (4 - V_t) \times 1 - \frac{1}{2} \times 1 \right] = \mu_n C_{ox} \frac{W}{L} (3.5 - V_t)$$

$$\Rightarrow \frac{I_{D1}}{I_{D2}} = \frac{72}{270} = \frac{(2 - V_t)^2}{2(3.5 - V_t)}$$

$$\Rightarrow V_t = 0.8\text{ V} \Rightarrow \mu_n C_{ox} \frac{W}{L} = 100\text{ }\mu\text{A/V}^2$$

### Problem 4.33

$$I_D = \frac{1}{2} k'_n \frac{W}{L} (V_{GS} - V_t)^2 \Rightarrow k'_n \frac{W}{L} = 1\text{ mA/V}^2$$

a).

$$V_1 = V_{DS} = 3\text{ V}$$

b).

$$V_2 = V_S = V_D - V_{DS} = 1 - 3 = -2\text{ V}$$

c).

$$V_3 = V_S = V_D - V_{DS} = 0 - (-3) = 3\text{ V}$$

d).

$$V_4 = V_D = V_S + V_{DS} = 5 - 3 = 2\text{ V}$$

When calculating  $R_{Dmax}$ ,  $V_{DS}$  has to be equal to  $V_{GS} - V_t$ , so that the device is operating on the edge of saturation:  $|V_{DS}| = 3 - 1 = 2\text{ V}$ .

a).

$$R_{Dmax} = \frac{3 - 2}{2\text{ mA}} = 0.5\text{ k}\Omega$$

b).

$$V_2 = -2\text{ V} \Rightarrow V_D = -2 + 3 = 1\text{ V} \Rightarrow R_{Dmax} = \frac{1}{2\text{ mA}} = 0.5\text{ k}\Omega$$

c).

$V_2 = -3\text{ V} \Rightarrow V_S = V_3 = 3\text{ V}$ . Now for  $V_{DS}$  to be  $-2\text{ V}$ ,  $V_D$  has to be  $1\text{ V}$ .

$$R_{Dmax} = \frac{1}{2\text{ mA}} = 0.5\text{ k}\Omega$$

d).

$$V_{GS} = -3 \text{ V} \Rightarrow V_G = V_4 = 2 \text{ V}$$

Adding the resistor between  $V_4$  and drain means that  $V_D$  has to be  $5 - 2 = 3 \text{ V}$  and this leaves  $1 \text{ V}$  to voltage drop on the resistor:

$$R_{Dmax} = \frac{1}{2 \text{ mA}} = 0.5 \text{ k}\Omega$$

When calculating  $R_{Smax}$ , since the gate doesn't draw any current, the value of the resistor is immaterial. Assuming that the voltage drop across the current source is at least  $2 \text{ V}$ ,  $R_{Smax}$  can be calculated as:

a).

$$V_1 = 8 \text{ V}, V_{GS} = 3 \text{ V} \Rightarrow V_S = 8 - 3 = 5 \text{ V}$$

$$\Rightarrow R_{Smax} = \frac{5}{2} = 2.5 \text{ k}\Omega$$

b).

$$V_2 = -9 + 2 = -7 \text{ V}, V_S = 1 - |V_{GS}| = -2 \text{ V}$$

$$\Rightarrow R_{Smax} = \frac{-2 - (-7)}{2} = 2.5 \text{ k}\Omega$$

c).

$$V_3 = 10 - 2 = 8 \text{ V}, V_S = 0 + |V_{GS}| = 3 \text{ V}$$

$$\Rightarrow R_{Smax} = \frac{8 - 3}{2} = 2.5 \text{ k}\Omega$$

d).

$$V_4 = -5 + 2 = -3 \text{ V}, V_S = -3 + |V_{GS}| = 0 \text{ V}$$

$$\Rightarrow R_{Smax} = \frac{5}{2} = 2.5 \text{ k}\Omega$$